

MEMORY DEVICES HAVING BIT LINE PRECHARGE CIRCUITS WITH OFF CURRENT PRECHARGE CONTROL AND ASSOCIATED BIT LINE PRECHARGE METHODS

Cross Reference to Related Application

This application claims priority under 35 U.S.C. § 119 from Korean Patent Application No. 2003-36748, filed June 9, 2003, the contents of which are incorporated herein by reference in their entirety.

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Field of the Invention

The present invention relates to semiconductor memory devices and, more particularly, to semiconductor memory devices having bit line precharge circuits and associated methods of pre-charging bit lines.

Background of the Invention

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As semiconductor memory processing technology has advanced, smaller and smaller semiconductor memory devices are being introduced. Deep sub-micron processes are now being developed to facilitate production of small, highly integrated, memory devices. In order to improve the performance of transistors developed using deep sub-micron technology, the threshold voltage (V_{th}) of the transistor may be reduced. When the threshold voltage is lowered, however, both the saturation current of the transistor in its "on" state and the leakage or "off-current" (I_{off}) of the transistor in its "off" state may increase.

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FIG. 1 is a diagram of a memory cell array **100** that may be used to illustrate the impact of a leakage current. The memory cell array includes a plurality of memory cells **102**, **104**, **106**, As will be appreciated by those of skill in the art, these individual memory cells **102**, **104**, **106** are typically arranged in rows and columns to form the memory cell array **100**. As shown in **FIG. 1**, in the memory cell

array **100**, the memory cells **102**, **104**, **106**, ... are located at intersections of a plurality of word lines (WL0, WL1, WL2, ...), a bit line (BL) and a complementary bit line (BLB). For convenience of explanation, in the following description it is assumed that the power supply voltage (VDD) corresponding to logic level "1" is stored in a first node (NA) and the ground or reference voltage (VSS) corresponding to logic level "0" is stored in a second node (NB). In the example of FIG. 1, the ground voltage (VSS) is set as 0 volts, although other ground voltages (VSS) may be used.

When the first word line (WL0) is enabled, the first memory cell **102** is activated and data stored in the first memory cell **102** is transferred to the bit line (BL) and the complementary bit line (BLB). Data in the first memory cell **102** experiences charge sharing such that a voltage difference between the bit line (BL) and the complementary bit line (BLB) occurs. Referring to FIG. 1, the bit line (BL) moves toward the power supply voltage (VDD) level and the complementary bit line (BLB) moves toward the ground voltage (VSS) level such that a voltage difference occurs. This voltage difference is sensed and amplified by a sense amplifier (not shown) so that the data in the memory cell **102** can be determined.

As shown in FIG. 1, both memory cell **104**, which is connected to the second word line (WL1), and memory cell **106**, which is connected to the third word line (WL2), are deactivated such that they are not connected to the bit line (BL) or the complementary bit line (BLB). However, the off-current (I_{off}) or leakage current of the memory cells **104** and **106** flows from the bit line (BL) into the transistors of memory cells **104** and **106** storing logic zeros such that the power supply voltage level of the bit line (BL) is lowered. As a result, the voltage difference between the bit line (BL) and the complementary bit line (BLB) decreases. This decrease results in a reduction of the sensing speed of the sense amplifier.

When the word lines (WL0, WL1, WL2, ...) are enabled in response to a decoded row address, the bit line (BL) and the complementary bit line (BLB) are precharged to the power supply voltage (VDD) in response to a precharge signal (PRE). After the bit line (BL) and the complimentary bit line (BLB) are precharged a voltage difference due to the memory cell data occurs that is sensed and amplified by the sense amplifier. Circuit blocks related to these operations are shown in FIG. 2.

As shown in FIG. 2, the circuit blocks include a memory cell array block **100**, a row decoder **210**, a pre-address decoding circuit and control signal generation unit

220, a bit line precharge unit 230 and a sense amplifier 240. The row decoder 210 and the pre-address decoding circuit 220 perform operations for enabling word lines (WL0, ..., WLn), and the bit line precharge circuit 230 that responds to a precharge signal (PRE) performs operations to precharge the bit line (BL) and the complementary bit line (BLB) to the power supply voltage (VDD) level. The sense amplifier 240 senses and amplifies the bit line (BL) and the complementary bit line (BLB) in which a voltage difference occurs in response to a control signal (SENSE).

FIG. 3 is a diagram illustrating the operational timing of the circuit of FIG. 2. Referring to FIG. 3, intervals (a) through (d) are shown. In interval (a), the precharge signal (PRE) is at a "low" logic level and in response to this the bit line (BL) and the complementary bit line (BLB) are precharged to the power supply voltage (VDD) level. Interval (b) is a floating interval during which the precharge signal (PRE) transitions to a "high" logic level (*i.e.*, is disabled). In interval (c), word line (WL0) transitions to a "high" logic level and is enabled. During interval (c) a voltage difference between the bit line (BL) and the complementary bit line (BLB) also occurs (due to the memory cell data), and a control signal (SENSE) transitions to a "high" logic level (*i.e.*, is enabled). As a result, the voltage difference between the bit line (BL) and the complementary bit line (BLB) is sensed and amplified. In interval (d), the precharge signal (PRE) transitions to a "low" logic level (*i.e.*, is enabled) such that the bit line (BL) and the complementary bit line (BLB) are again precharged to the power supply voltage (VDD) level.

In the operation of the circuit of FIG. 2, during interval (b), before the word line (WL0) is enabled, the voltage levels of the bit line (BL) and the complementary bit line (BLB), which are precharged to the power supply voltage (VDD) level, are lowered due to the effect of the off-current (*I_{off}*) described above with respect to FIG. 1. Accordingly, additional time may be required to arrive at the voltage difference between the bit line (BL) and the complementary bit line (BLB) that can be sensed by the sense amplifier. Thus, the off-current can act to reduce the operation speed of the memory device.

Summary of the Invention

Embodiments of the present invention provide semiconductor memory devices that have a memory cell array, an address decoder, a precharge control circuit and a precharge unit. The memory cell array has a set of memory cells that are accessed via

a set of word lines and first and second bit lines. The address decoder, which is coupled to the word lines, decodes a received address signal. The precharge control circuit generates a precharge signal in response to a precharge enable signal and a precharge delay signal, and the precharge unit precharges the first and second bit lines
5 in response to the precharge signal. In some embodiments of the present invention, the precharge control circuit may generate the precharge signal by performing a logical AND operation on the precharge enable signal and the precharge delay signal.

The semiconductor memory device may also include a delay circuit that generates the precharge delay signal by delaying the precharge enable signal for a
10 predetermined delay time. The delay circuit may be embodied, for example, as a NOR gate which receives the precharge enable signal and an inverter which inverts the output of the NOR gate. The predetermined delay time may be the time that it takes the word lines to become enabled in response to a transition of the decoded address signal. In embodiments of the present invention, the precharge control circuit
15 may be implemented as a NAND gate which receives the precharge enable signal and the precharge delay signal and an inverter which inverts the output of the NAND gate.

The precharge unit may be embodied as first and second transistors that precharge the first bit line and the second bit line, respectively, to the power supply voltage level in response to the precharge signal and a third transistor which equalizes
20 the voltage of the first bit line and the second bit line. The first, second and third transistors may be PMOS transistors. The precharge signal may be disabled after one of the plurality of word lines is enabled in response to the decoded address signal. This disablement of the precharge signal may occur a predetermined time after the precharge enable signal is disabled. The precharge signal may be enabled at
25 substantially the same time that the precharge enable signal is enabled.

Pursuant to further embodiments of the present invention, the precharge control circuit of the semiconductor memory devices may generate the precharge signal in response to the decoded address signal and the precharge enable signal. In these embodiments, the precharge unit precharges the first and second bit lines in
30 response to the precharge signal, and the precharge signal is disabled after one of the word lines is enabled. In these embodiments, the precharge control circuit may include a NOR gate which receives the decoded address signals, a first inverter which inverts the output of the NOR gate, a NAND gate which receives the output of the

first inverter and the precharge enable signal and a second inverter which inverts the output of the NAND gate to generate the precharge signal.

Pursuant to further embodiments of the present invention, methods of pre-charging first and second bit lines on a semiconductor memory device are provided.

5 Pursuant to these methods, the first and second bit lines are precharged in response to a precharge enable signal transitioning to a first level. This pre-charging operation continues until a predetermined time after the precharge enable signal transitions to a second level. The predetermined time may be the time that it takes the word lines to become enabled in response to a transition of the decoded address signal. The
10 precharge of the first and second bit lines may be resumed in response to the precharge enable signal transitioning back to the first level.

In still further embodiments of the present invention, methods for pre-charging a first bit line and a second bit line of a memory cell array are provided in which a precharge signal is generated in response to a precharge enable signal and a precharge
15 delay signal. The first and second bit lines are then precharged in response to the precharge signal. A word line is enabled in response to a decoded address signal. The precharge signal is then disabled after the word line is enabled. In these methods, the precharge delay signal may be generated by delaying the precharge enable signal for a predetermined time, and the predetermined delay time may be the time that it
20 takes the word lines to become enabled in response to a transition of the decoded address signal.

In embodiments of the present invention the precharge signal is only disabled after the word line is enabled. As a result, the effect of the off-current on the circuit may be reduced or eliminated, and the voltage difference between the bit line and the
25 complementary bit line may be increased providing an enhanced sensing margin.

Brief Description of the Drawings

FIG. 1 is a diagram of a prior art memory cell array;

FIG. 2 is a diagram illustrating a circuit included in a prior art memory device;

30 **FIG. 3** is a timing diagram illustrating the operational timing of the memory device of **FIG. 2**;

FIG. 4 is a diagram illustrating a circuit and method of generating precharge signals according to embodiments of the present invention;

FIG. 5 is a diagram of a memory device according to embodiments of the present invention;

FIG. 6 is a timing diagram illustrating the operational timing of the memory device of **FIG. 5**;

5 **FIG. 7** is a pair of graphs illustrating the simulated performance of memory devices according to embodiments of the present invention; and

FIG. 8 is a pair of graphs illustrating the simulated performance of the prior art memory device of **FIG. 1**.

Detailed Description

10 The present invention will now be described more fully with reference to the accompanying drawings, in which typical embodiments of the invention are shown. This invention, however, may be embodied in many different forms and should not be construed as limited to the embodiments set forth herein. Rather, these embodiments are provided so that this disclosure will be thorough and complete, and will fully
15 convey the scope of the invention to those skilled in the art. Moreover, each embodiment described and illustrated herein includes its complementary conductivity type embodiment as well. Like reference numerals refer to like elements throughout.

FIG. 4 depicts a circuit and method for generating a precharge signal (PRE). The precharge signal (PRE) is activated and inactivated in response to a precharge enable signal (PRE_EN). As shown in **FIG. 4**, the precharge enable signal (PRE_EN)
20 is input to a delay circuit **410** which delays the precharge enable signal (PRE_EN) for a predetermined time (ΔT). Both the delayed precharge enable signal that is output from the delay circuit **410** and the precharge enable signal (PRE_EN) are input to a NAND gate **420**. The output of the NAND gate **420** is input to an inverter **430** which
25 outputs the precharge signal (PRE). Thus, according to the precharge control circuit of **FIG. 4**, the precharge signal (PRE) is inactivated a predetermined time (ΔT) after the precharge enable signal (PRE_EN) is inactivated, while the precharge signal (PRE) is activated at the same time that the precharge enable signal (PRE_EN) is activated.

30 Embodiments of the present invention and their operation will now be explained and contrasted to the operation of the precharge circuit **230** of the prior art memory device of **FIG. 2**.

The precharge circuit **230** of **FIG. 2** comprises first and second PMOS transistors **232** and **234**, which precharge the bit line (BL) and the complementary bit line (BLB) to the power supply voltage (VDD) level in response to the precharge signal (PRE). The precharge circuit **230** further includes a third PMOS transistor **236** which equalizes the bit line (BL) and the complementary bit line (BLB) to the power supply voltage (VDD) level in response to the precharge signal (PRE). The "precharge signal" refers to any signal that either directly or indirectly acts to cause the precharge circuit **230** to precharge one or more bit lines. The precharge circuit **230** precharges the bit line (BL) and the complementary bit line (BLB) to the power supply voltage (VDD) level when the precharge signal (PRE) is at a "low" logic level, and disables the precharge operation of the bit line (BL) and the complementary bit line (BLB) when the precharge signal (PRE) is at a "high" logic level. It will be appreciated by those of skill in the art that the precharge circuit can be implemented in a wide variety of different ways, and that the present invention is not limited to the exemplary embodiments depicted in **FIGS. 2** and **5**, but encompasses any precharge unit that is capable of pre-charging the bit line (BL) and the complimentary bit line (BLB).

In the circuit depicted in **FIG. 4**, when the precharge enable signal (PRE_EN) is at a "low" logic level, thereby initiating a bit line precharge operation, the output of the NAND gate **420** is at "high" logic level. The inverter **430** converts this "high" logic signal to a "low" logic signal that is output as the precharge signal (PRE). As noted above, when the precharge signal (PRE) is at a "low" logic level, the bit line (BL) and the complementary bit line (BLB) are precharged to the power supply voltage (VDD) level. A predetermined time (ΔT), which corresponds to the delay time of the delay circuit **410**, after the time when the precharge enable signal (PRE_EN) transitions to a "high" logic level, the output of the NAND gate **420** transitions to a "low" logic level signal and the precharge signal (PRE) transitions to a "high" logic level. As illustrated in **FIG. 2**, the transition of the precharge signal (PRE) to a "high" logic level causes the PMOS transistors **232**, **234** and **236** of the precharge circuit **230** to turn off and the precharge operation of the bit line (BL) and the complementary bit line (BLB) is disabled.

In embodiments of the present invention, the predetermined time (ΔT) may be the time that it takes to enable a word line in response to decoded row addresses (*i.e.*, the delay between the time when the row addresses transition to the time at which the

word line is enabled in response thereto). When the predetermined time (ΔT) is set in such a manner the precharge operation will be disabled after the word line is enabled.

FIG. 5 is a circuit diagram illustrating a memory device according to embodiments of the present invention that includes a precharge control circuit. As shown in **FIG. 5**, the memory device **500**, like the memory device **200** of **FIG. 2**, may include a memory cell array block **100**, a row decoder **210**, a pre-address decoding circuit and control signal generation unit **220**, a bit line precharge unit **230**, and a sense amplifier **240**. Additionally, the memory device **500** of **FIG. 5** further includes a precharge control circuit unit **510**.

As shown in **FIG. 5**, in embodiments of the present invention the precharge control circuit unit **510** may be implemented as a NOR gate **512** which receives decoded row addresses provided by the pre-address decoding circuit **220**, a first inverter **514** which receives the output of the NOR gate **512**, a NAND gate **516** which receives the output of the first inverter **514** and a precharge enable signal (PRE_EN), and a second inverter **518** which inverts the output of the NAND gate **516** and generates a precharge signal (PRE).

In response to the transition of the decoded row addresses to a "high" logic level, the NOR gate **512** outputs a "low" logic level signal. When this occurs, if the precharge enable signal (PRE_EN) is at a "low" logic level, the NAND gate **516** and the inverter **518** generate a precharge signal (PRE) that is at a "low" logic level and the bit line (BL) and the complementary bit line (BLB) are precharged. If the precharge enable signal (PRE_EN) is at a "high" logic level, the NAND gate **516** and the inverter **518** generate a precharge signal (PRE) that is at a "high" logic level and the precharge of the bit line (BL) and the complementary bit line (BLB) is disabled.

If instead the decoded row addresses are at a "low" logic level, the NOR gate **512** outputs a "high" logic level signal and the NAND gate **516** and the inverter **518** generate a precharge signal (PRE) that is at a "low" logic level regardless of the logic level of the precharge enable signal (PRE_EN). In this manner, the bit line (BL) and the complementary bit line (BLB) are precharged during a time when the word lines (WL0, ..., WLn) are not enabled.

FIG. 6 is a timing diagram illustrating the operational timing of the memory device of **FIG. 5**. A comparison of **FIG. 6** to the timing diagram of **FIG. 3** shows that the timing diagram of **FIG. 6** includes intervals (a), (c) and (d) but does not include the interval (b) that is present in the timing diagram of **FIG. 3**. In the timing

diagram of **FIG. 6**, during interval (a) the bit line (BL) and the complementary bit line (BLB) are precharged to the power supply voltage (VDD) level in response to the precharge signal (PRE) being in a "low" logic level. In interval (c), the precharge signal (PRE) transitions to a "high" logic level (*i.e.*, is disabled) in response to word line (WL0) being enabled to a "high" logic level. According to the data of the memory cell connected to the enabled word line (WL0), the bit line (BL) and the complementary bit line (BLB) experience charge sharing and the voltage difference between the bit line (BL) and the complementary bit line (BLB) increases. The sensing enable signal (SENSE) senses this voltage difference between the bit line (BL) and the complementary bit line (BLB). During interval (d), word line (WL0) is disabled and the precharge signal (PRE) returns to a "low" logic level. In response to this transition of the precharge signal (PRE), the bit line (BL) and the complementary bit line (BLB) are again precharged.

As discussed above, the precharge control circuit **510** may include circuit elements that implement the predetermined delay that is applied to the precharge enable signal in embodiments of the present invention. In other embodiments of the present invention, the semiconductor memory device may include a separate delay circuit that generates a precharge delay signal by, for example, delaying the precharge enable signal for a predetermined time. The precharge delay signal may then be input to the precharge control circuit, which in response to the precharge delay signal and the precharge enable signal generates the precharge signal.

FIG. 7 is a diagram illustrating simulated performance of memory devices according to embodiments of the present invention. As shown in **FIG. 7**, in response to the decoded address signal, word line (WL) is enabled and transitions to a "high" logic level. After a predetermined time (ΔT), the precharge signal (PRE) transitions to a "high" logic level and is disabled. Starting at the time when the word line (WL) is enabled, the voltage difference between the bit line (BL) and the complementary bit line (BLB) is slowly generated, and at the time when the precharge signal (PRE) is disabled, the voltage difference between the bit line (BL) and the complementary bit line (BLB) increases further. Since the voltage difference between the bit line (BL) and the complementary bit line (BLB) is relatively large, the sensing margin can be made greater during an activated interval of the sensing enable signal (SENSE) where the voltage difference between the bit line (BL) and the complementary bit line (BLB) is sensed and amplified.

FIG. 8 is a diagram illustrating the simulated performance of the prior art memory device of **FIG. 2**. As shown in **FIG. 8**, the precharge signal (PRE) transitions to a "high" logic level and is disabled. After a predetermined time (ΔT), the word line (WL) transitions to a "high" logic level and is enabled. Once the precharge signal (PRE) is disabled, the voltage difference between the bit line (BL) and the complementary bit line (BLB) begins to take place. However, due to the effect of the off-current (I_{off}), the bit line moves toward a "low" logic level and the complementary bit line (BLB) moves toward a "high" logic level, which is the opposite of expected normal voltage levels. After the word line (WL) is enabled and a subsequent delay transpires, the direction of the bit line (BL) movement is reversed toward the "high" logic level and the direction of the complementary bit line (BLB) movement is reversed toward the "low" logic level and the voltage difference increases. However, as a result of the off-current, the voltage difference between the bit line (BL) and the complementary bit line (BLB) may be relatively small. Thus, during the activated interval of the sensing enable signal (SENSE) where the voltage difference between the bit line (BL) and the complementary bit line (BLB) is sensed and amplified, the sensing margin decreases.

The methods and systems for enabling the word line (WL) and then disabling the precharge signal (PRE) according to embodiments of the present invention may cause current consumption during the time interval between the enabling of the word line (WL) to the disabling of the precharge signal (PRE) due to formation of a current path between the power supply voltage that is the bit line precharge voltage and memory cell data that is in a "low" logic level. However, the methods and systems according to embodiments of the present invention can increase the voltage difference between the bit line (BL) and the complementary bit line (BLB) while reducing or eliminating the effects of the off-current (I_{off}).

While this invention has been particularly shown and described with reference to certain embodiments thereof, it will be understood by those skilled in the art that various changes in form and details may be made therein without departing from the spirit and scope of the invention as defined by the appended claims and equivalents.